

LETTER

# Behavior of intercore crosstalk in square-layout uncoupled four-core fibers

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**Abstract** Recently, a very interesting behavior of crosstalk (XT) between diagonal cores (diagonal XT) in a square-layout uncoupled four-core fiber (UC-4CF) that grows quadratically with fiber length has been reported. In this paper, the behavior of intercore XT in square-layout UC-4CFs is investigated on the basis of a coupled-power theory. To make it easy to understand the physical mechanism of intercore XT behavior, we derive an approximate solution of coupled-power equations. We show that there exists a range of fiber lengths within which the diagonal XT grows quadratically. Finally, we propose a heterogeneous square-layout UC-4CF in which intercore XT could be further suppressed.

**Keywords:** four-core fiber, intercore crosstalk, coupled-power theory

**Classification:** Optical hardware (fiber optics, microwave photonics, optical interconnects, photonic signal processing, photonic integration and modules, optical sensing, etc.)

## 1. Introduction

Uncoupled multicore fibers (MCFs) with the standard 125- $\mu\text{m}$  cladding have attracted attention because conventional cabling and splicing technologies can be effectively adopted for such fibers: two-core fibers [1], four-core fibers [2, 3, 4, 5, 6], five-core fibers [7, 8], and eight-core fibers [9, 10]. Recently, high-capacity and/or long-haul transmission experiments have been successfully performed using a square-layout uncoupled four-core fiber (UC-4CF) with a standard cladding diameter [11, 12, 13, 14]. The square-layout UC-4CF is also suitable for bidirectional transmissions [15] in which intercore crosstalk (XT) is suppressed by assigning opposite signal propagation directions between adjacent cores [16, 17, 18, 19]. XT in bidirectional transmissions is predominantly induced by Rayleigh backscattered XT ( $\text{XT}_{\text{bs}}$ ) and back-reflected XT ( $\text{XT}_{\text{refl}}$ ) from counterpropagating signals in the nearest neighboring cores, as well as diagonal XT ( $\text{XT}_{\text{diag}}$ ), i.e., XT between diagonal cores [15]. In bidirectional transmissions with square-layout UC-4CFs, care should be taken for not only  $\text{XT}_{\text{bs}}$  and  $\text{XT}_{\text{refl}}$  but also  $\text{XT}_{\text{diag}}$  because signals in diagonal cores are transmitted in the same direction.

In general, adjacent XT ( $\text{XT}_{\text{adj}}$ ), i.e., intercore XT between adjacent cores, grows linearly with fiber length [20, 21, 22, 23]. More recently, Nakagawa *et al.* [24] have measured intercore XT in a square-layout UC-4CF and found

a very interesting behavior of  $\text{XT}_{\text{diag}}$  that grows quadratically with fiber length.

In this paper, the behavior of intercore XT in square-layout UC-4CFs is investigated on the basis of a coupled-power theory [20, 21, 22, 23]. To make it easy to understand the physical mechanism of intercore XT behavior, we derive an approximate solution of coupled-power equations and confirm that  $\text{XT}_{\text{diag}}$  is induced by direct XT ( $\text{XT}_{\text{dir}}$ ) from a diagonal core and indirect XT ( $\text{XT}_{\text{indir}}$ ) via adjacent cores.  $\text{XT}_{\text{dir}}$  and  $\text{XT}_{\text{indir}}$  grow linearly and quadratically with fiber length, respectively. We show that  $\text{XT}_{\text{diag}}$  has a specific fiber length at which  $\text{XT}_{\text{dir}}$  and  $\text{XT}_{\text{indir}}$  are equal to each other. This length gives us a lower bound of fiber length beyond which  $\text{XT}_{\text{indir}}$  becomes predominant,  $\text{XT}_{\text{diag}} \approx \text{XT}_{\text{indir}}$ . We also show that there is an upper bound of fiber length below which  $\text{XT}_{\text{diag}}$  grows quadratically. As a result, we conclude that whether  $\text{XT}_{\text{diag}}$  grows linearly or quadratically depends on fiber length. Finally, to further suppress not only  $\text{XT}_{\text{diag}}$  but also  $\text{XT}_{\text{bs}}$  and  $\text{XT}_{\text{refl}}$ , we propose a square-layout UC-4CF with heterogeneous cores.

## 2. Solutions of coupled-power equations

We consider the square-layout UC-4CF with homogeneous cores shown in Fig. 1(a). Fig. 1(b) shows a cross section of the fabricated step-index profile UC-4CF with the standard 125- $\mu\text{m}$  cladding [24].

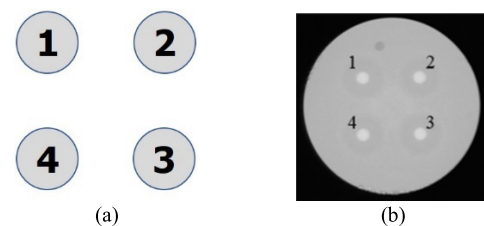
Assuming that all cores are homogeneous (identical), coupled-power equations are given by [24, 25]

$$\frac{dP_1(z)}{dz} = h(P_2 - P_1) + g(P_3 - P_1) + h(P_4 - P_1), \quad (1a)$$

$$\frac{dP_2(z)}{dz} = h(P_1 - P_2) + h(P_3 - P_2) + g(P_4 - P_2), \quad (1b)$$

$$\frac{dP_3(z)}{dz} = g(P_1 - P_3) + h(P_2 - P_3) + h(P_4 - P_3), \quad (1c)$$

$$\frac{dP_4(z)}{dz} = h(P_1 - P_4) + g(P_2 - P_4) + h(P_3 - P_4), \quad (1d)$$



**Fig. 1** (a) Schematic and (b) cross section [24] of a square-layout UC-4CF with homogeneous cores.

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**Table I** Analytical expressions for intercore crosstalk in square-layout UC-4CFs with homogeneous cores: exact crosstalk [24, 25] and approximate crosstalk (this work).

$g$	XT	Exact XT	Approx. XT
$g \neq 0$	XT <sub>adj</sub>	$\frac{1 - e^{-4hL}}{1 + 2e^{-2(h+g)L} + e^{-4hL}}$	$hL$
	XT <sub>diag</sub>	$\frac{1 - 2e^{-2(h+g)L} + e^{-4hL}}{1 + 2e^{-2(h+g)L} + e^{-4hL}}$	$gL + h^2L^2$
$g = 0$	XT <sub>adj</sub>	$\tanh(hL)$	$hL$
	XT <sub>diag</sub>	$\tanh^2(hL)$	$h^2L^2$

where  $P_i(z)$  is the optical power in core  $i$  ( $i = 1, 2, 3, 4$ ) and  $z$  represents the propagation direction.  $h$  and  $g$  are the average power-coupling coefficients (PCCs) [22, 23] between adjacent cores and between diagonal cores, respectively, and are assumed to be invariant along the  $z$ -axis.

The exact solutions of Eqs. (1a)–(1d) with and without  $g$  have been derived by Liang *et al.* [25] and Nakagawa *et al.* [24], respectively. In Refs. [24, 25], only the results are presented; thus, we show the procedure for deriving the exact solutions in Appendix.

When core  $i$  is excited with incident power  $P_{in}$ ,  $P_i(z = 0) = P_{in}$ , the exact solutions [24, 25] satisfy the power conservation relation as

$$P_i(z) + \sum_{j \neq i} P_j(z) = P_{in}. \quad (2)$$

When  $z = \infty$ , the power is uniformly distributed as  $P_1 = P_2 = P_3 = P_4 = P_{in}/4$ . The intercore XT from core  $i$  to core  $j$  ( $j \neq i$ ) at  $z = L$  is defined as

$$XT_{ji} = \frac{P_j(L)}{P_i(L)}, \quad (3)$$

where  $L$  is the fiber length. Analytical expressions for intercore XT in square-layout UC-4CFs that are deduced from the exact solutions of Eqs. (1a)–(1d) [24, 25] are shown in Table I.

Although exact solutions of coupled-power equations are indispensable to the highly accurate evaluation of intercore XT in MCFs, it would be difficult to give a physical interpretation of intercore XT behavior. Thus, to make it easy to understand the physical mechanism of intercore XT behavior in square-layout UC-4CFs, we attempt to derive an approximate solution of Eqs. (1a)–(1d) that is valid for small XT values.

When core 1 is excited with incident power  $P_{in}$ ,  $P_1(z = 0) = P_{in}$ , on the basis of Ref. [10] in which a three-core model is introduced, we make the following assumptions:

$$P_1(z) \gg P_2(z) \gg P_3(z), \quad (4a)$$

$$P_1(z) \gg P_4(z) \gg P_3(z), \quad (4b)$$

$$g \ll h \ll 1, \quad (4c)$$

where we note that an average PCC exponentially decreases with increasing core-to-core distance [10] and that in square-layout UC-4CFs, the distance between diagonal cores is  $\sqrt{2}$  times that between adjacent cores; therefore,  $g$  is sufficiently

smaller than  $h$ . Considering the assumptions in Eqs. (4a)–(4c), Eqs. (1a)–(1d) are reduced to

$$P_1(z) \approx P_{in}, \quad (5a)$$

$$\frac{dP_2(z)}{dz} = \frac{dP_4(z)}{dz} \approx hP_1(z), \quad (5b)$$

$$\frac{dP_3(z)}{dz} \approx gP_1(z) + hP_2(z) + hP_4(z), \quad (5c)$$

where owing to the very small XT, we also assume that the optical power in core 1 is not depleted.

Substituting Eq. (5a) into Eq. (5b) and noting  $P_2(0) = P_4(0) = 0$ , we obtain  $P_2(z)$  and  $P_4(z)$  as

$$P_2(z) = P_4(z) \approx hzP_{in}. \quad (6)$$

Similarly, substituting Eqs. (5a) and (6) into Eq. (5c) and noting  $P_3(0) = 0$ , we obtain  $P_3(z)$  as

$$P_3(z) \approx (gz + h^2z^2)P_{in}. \quad (7)$$

From Eqs. (6) and (7), we have the approximate XT describing the fiber length dependence of adjacent XT ( $XT_{21} = XT_{41} \equiv XT_{adj}$ ) and diagonal XT ( $XT_{31} \equiv XT_{diag}$ ), as shown in Table I. As is well known,  $XT_{adj}$  is given by the product of  $h$  and  $L$  [22]. On the other hand,  $XT_{diag}$  is given by the sum of  $XT_{dir} = gL$  and  $XT_{indir} = h^2L^2$ . In the three-core model,  $XT_{indir} = h^2L^2/2$  [10]. This is due to the fact that  $XT_{indir}$  occurs via one neighboring core in the three-core model and via two neighboring cores in the present square-layout four-core model.

From Table I, we can see that  $XT_{adj}$  grows linearly with increasing fiber length. However, this is not the case in  $XT_{diag}$ .  $XT_{diag}$  has a specific fiber length at which  $XT_{dir}$  and  $XT_{indir}$  are equal to each other and this length is expressed as  $L_{ip} = g/h^2$  (intersecting point of  $XT_{dir}$  and  $XT_{indir}$  lines).  $XT_{diag}$  is predominantly induced by  $XT_{dir}$  under  $L < L_{ip}$  and by  $XT_{indir}$  under  $L > L_{ip}$ . That is,  $L_{ip}$  gives us a lower bound of fiber length beyond which  $XT_{diag}$  grows quadratically. Note that such behavior in  $XT_{adj}$  and  $XT_{diag}$  does not last indefinitely. The reason is that as the fiber length increases, the accuracy of the approximate solution decreases.

Thus, using exact solutions, we define an application range of the approximate solution as follows:

$$|XT_{approx} - XT_{exact}| \leq 0.1 \text{ dB}, \quad (8)$$

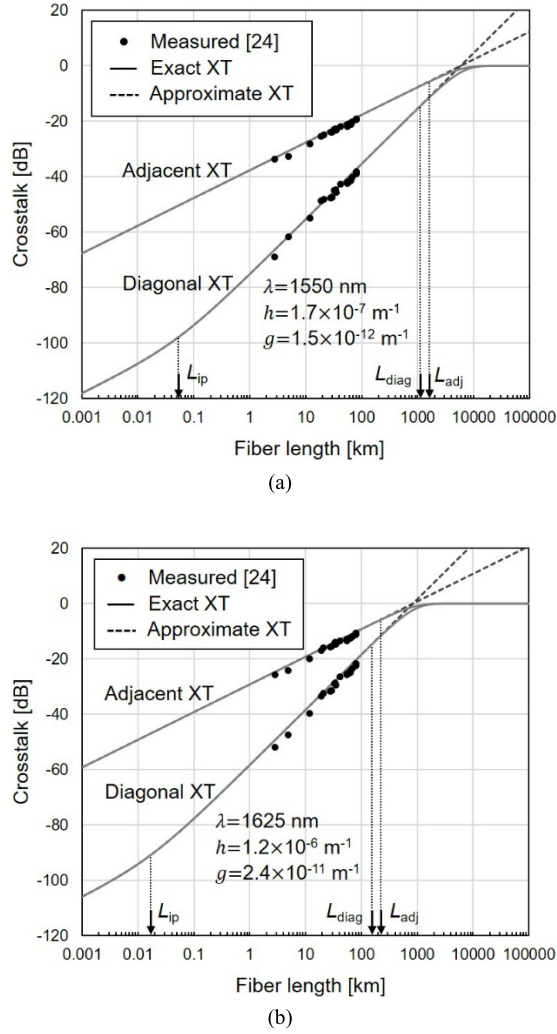
where  $XT_{approx}$  is the approximate XT and  $XT_{exact}$  is the exact XT.

Using the exact and approximate XTs in Table I and noting that when  $XT_{indir}$  is predominant ( $XT_{diag} \approx XT_{indir}$ ),  $XT_{dir}$  is negligibly small (we can use the approximation  $g \approx 0$ ), we obtain the maximum values of fiber length satisfying Eq. (8), namely,  $L_{adj}$  for  $XT_{adj}$  and  $L_{diag}$  for  $XT_{diag}$ , by solving

$$\frac{hL_{adj}}{\tanh(hL_{adj})} = 10^{0.01}, \quad (9a)$$

$$\frac{hL_{diag}}{\tanh(hL_{diag})} = 10^{0.005}, \quad (9b)$$

where we note that in this case, the approximate XT is always greater than the exact XT.



**Fig. 2** Fiber length dependence of intercore crosstalk in the square-layout UC-4CF with homogeneous cores at the wavelength ( $\lambda$ ) of (a) 1550 nm and (b) 1625 nm.

From Eqs. (9a) and (9b), we obtain  $L_{adj} \approx 0.265/h$  and  $L_{diag} \approx 0.187/h$ . As a result, we can say that  $XT_{adj}$  grows linearly under  $L < L_{adj}$  and that  $XT_{diag}$  grows linearly under  $L < L_{ip}$  and quadratically under  $L_{ip} < L < L_{diag}$ .

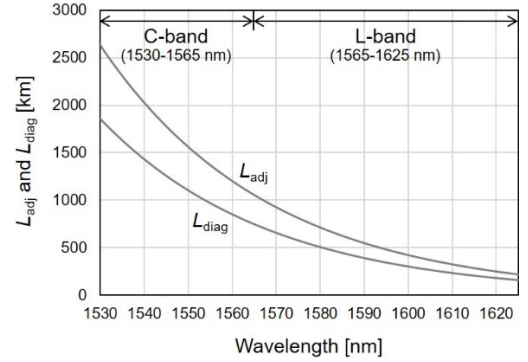
### 3. Results and discussion

Here, we consider the square-layout UC-4CF shown in Fig. 1(b) [24]. The average distance between adjacent cores is  $39.7 \mu\text{m}$  and that between diagonal cores is  $56.1 \mu\text{m}$  [24]. The bending radius is assumed to be  $155 \text{ mm}$ . Using Eq. (25) in Ref. [22], the values of  $h$  and  $g$  are evaluated to be  $1.7 \times 10^{-7} \text{ m}^{-1}$  and  $1.5 \times 10^{-12} \text{ m}^{-1}$  at 1550 nm (C-band), respectively, and  $1.2 \times 10^{-6} \text{ m}^{-1}$  and  $2.4 \times 10^{-11} \text{ m}^{-1}$  at 1625 nm (L-band), respectively. Since the average PCCs ( $h$  and  $g$ ) involve the mode-coupling coefficient [22, 23] which is expressed in terms of the overlap integral of electromagnetic fields [26], the overlap integral was numerically calculated.

Figs. 2(a) and (b) show the fiber length dependence of intercore XT values at 1550 nm and 1625 nm, respectively, where  $g \neq 0$ . Table II shows the values of  $L_{ip}$ ,  $L_{diag}$ , and  $L_{adj}$ . From Figs. 2(a) and (b), we can see that the calculated

**Table II**  $L_{ip}$ ,  $L_{diag}$ , and  $L_{adj}$ .

Wavelength	$L_{ip}$	$L_{diag}$	$L_{adj}$
1550 nm	$\sim 52 \text{ m}$	$\sim 1100 \text{ km}$	$\sim 1560 \text{ km}$
1625 nm	$\sim 17 \text{ m}$	$\sim 160 \text{ km}$	$\sim 220 \text{ km}$



**Fig. 3** Wavelength dependence of  $L_{adj}$  and  $L_{diag}$ .

results obtained using the exact-XT expression (solid lines) and approximate-XT expression (dashed lines) agree well with the measured results [24] denoted by dots. Note that the accuracy of the approximate solution decreases under  $L > L_{adj}$  and  $L > L_{diag}$ .

In Ref. [24], the intercore XT was measured for fiber lengths ranging from 3 km to 80 km. Since the fiber length satisfies the conditions  $L < L_{diag}$  and  $L < L_{adj}$ , the intercore XT behavior can be physically explained with the approximate-XT expression derived here.  $XT_{diag}$  measurement in Ref. [24] was performed under  $L_{ip} < L < L_{diag}$  within which  $XT_{indir}$  is predominant; therefore, it follows that  $XT_{diag}$  grows quadratically with fiber length [24]. Note that  $XT_{diag}$  does not grow quadratically under  $L < L_{ip}$  and  $L > L_{diag}$ , and that  $XT_{adj}$  does not grow linearly under  $L > L_{adj}$ .

Fig. 3 shows the wavelength dependence of  $L_{adj}$  and  $L_{diag}$  in the C-band (1530–1565 nm) and L-band (1565–1625 nm).  $L_{adj}$  and  $L_{diag}$  decrease exponentially with increasing wavelength.

### 4. Proposal of heterogeneous square-layout UC-4CFs

Thus far, we have discussed square-layout UC-4CFs with homogeneous cores. To further suppress not only  $XT_{diag}$  but also  $XT_{bs}$  and  $XT_{refl}$ , we propose the square-layout UC-4CF with heterogeneous cores shown in Fig. 4, where cores 2 and 4 are different from cores 1 and 3. In heterogeneous MCFs [27], not only identical cores but also non-identical cores are arranged so that the intercore XT between any pair of adjacent cores becomes sufficiently small; therefore, cores are more closely packed in a definite space than in a conventional, homogeneous MCF. Furthermore, the close packing of cores helps increase the outer cladding thickness (the minimum distance between the outermost core center and the cladding-coating interface), resulting in the suppression of coating-leakage loss (tunneling loss from the core to the fiber coating). Also, the XT relaxation can be used for the design of larger effective area which mitigates nonlinear

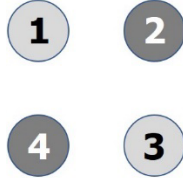


Fig. 4 Schematic of a square-layout UC-4CF with heterogeneous cores.

effects in optical fibers and tighter core pitch which is advantageous to improve the rotation misalignment tolerance at the connection point [15].

For the heterogeneous UC-4CF shown in Fig. 4, coupled-power equations are given by

$$\frac{dP_1(z)}{dz} = \tilde{h}(P_2 - P_1) + g_1(P_3 - P_1) + \tilde{h}(P_4 - P_1), \quad (10a)$$

$$\frac{dP_2(z)}{dz} = \tilde{h}(P_1 - P_2) + \tilde{h}(P_3 - P_2) + g_2(P_4 - P_2), \quad (10b)$$

$$\frac{dP_3(z)}{dz} = g_1(P_1 - P_3) + \tilde{h}(P_2 - P_3) + \tilde{h}(P_4 - P_3), \quad (10c)$$

$$\frac{dP_4(z)}{dz} = \tilde{h}(P_1 - P_4) + g_2(P_2 - P_4) + \tilde{h}(P_3 - P_4), \quad (10d)$$

where  $\tilde{h}$  is the average PCC between adjacent non-identical cores, and  $g_1$  and  $g_2$  are the average PCCs between identical cores 1 and 3 and between identical cores 2 and 4, respectively.

Analytical expressions for the intercore XT in square-layout UC-4CFs with heterogeneous cores are similar to those in Table I. When core 1 or 3 is excited at the fiber input end,  $h$  and  $g$  are replaced by  $\tilde{h}$  and  $g_1$ , respectively. When core 2 or 4 is excited at the fiber input end, on the other hand,  $h$  and  $g$  are replaced by  $\tilde{h}$  and  $g_2$ , respectively. When the distance between adjacent cores of the heterogeneous UC-4CF is the same as that of the homogeneous one, the relation  $\tilde{h} < h$  holds in the non-phase matching region where the bending radius is larger than the so-called critical (or threshold) bending radius [21, 22, 23, 28, 29]; thus, in the heterogeneous UC-4CFs, not only  $XT_{\text{diag}}$  but also  $XT_{\text{bs}}$  and  $XT_{\text{refl}}$  could be further suppressed ( $XT_{\text{bs}}$  and  $XT_{\text{refl}}$  are proportional to the average PCC between adjacent cores [15]).

When  $\tilde{h} \approx g_l$  ( $l = 1, 2$ ) or  $\tilde{h} < g_l$ , we note that the assumptions in Eqs. (4a)–(4c) made in deriving Eqs. (5a)–(5c) do not hold and that it is necessary to rewrite the analytical expressions for the intercore XT in homogeneous UC-4CFs shown in Table I.

As an example, when  $\tilde{h} \approx g_l$ , from Table I we obtain the exact XT as

$$XT_{\text{adj}} = XT_{\text{diag}} = \frac{1 - e^{-4\tilde{h}L}}{1 + 3e^{-4\tilde{h}L}}, \quad (11)$$

and the approximate XT as

$$XT_{\text{adj}} = XT_{\text{diag}} = \tilde{h}L. \quad (12)$$

The maximum fiber length satisfying Eq. (8),  $L_{\text{adj}}$  for  $XT_{\text{adj}}$ , is obtained by solving

$$\frac{\tilde{h}L_{\text{adj}}(1 + 3e^{-4\tilde{h}L_{\text{adj}}})}{1 - e^{-4\tilde{h}L_{\text{adj}}}} = 10^{-0.01}, \quad (13)$$

where we note that in the small XT region, the approximate XT is less than the exact XT. Obviously,  $L_{\text{diag}}$  for  $XT_{\text{diag}}$  is equal to  $L_{\text{adj}}$ . From Eq. (13), we have  $L_{\text{adj}} = L_{\text{diag}} \approx 0.0235/\tilde{h}$ . As a result, we can say that both  $XT_{\text{adj}}$  and  $XT_{\text{diag}}$  grow linearly under  $L < L_{\text{adj}} = L_{\text{diag}}$ . In this case, i.e.,  $\tilde{h} \approx g_l \ll h$ , the lengths denoted by  $L_{\text{adj}}$  and  $L_{\text{diag}}$  in the heterogeneous UC-4CF become markedly higher than those in the homogeneous one.

## 5. Conclusion

The behavior of intercore XT in square-layout UC-4CFs was investigated on the basis of a coupled-power theory. To make it easy to understand the physical mechanism of intercore XT behavior, we derived an approximate solution of coupled-power equations. We showed that there exists a range of fiber lengths within which  $XT_{\text{diag}}$  grows quadratically. Finally, we proposed a heterogeneous square-layout UC-4CF and suggested that not only  $XT_{\text{diag}}$  but also  $XT_{\text{bs}}$  and  $XT_{\text{refl}}$  could be further suppressed.

A square-layout UC-4CF with standard 125- $\mu\text{m}$  cladding is also suitable for O-band (1260–1360 nm) short-reach transmission. Other types of 125- $\mu\text{m}$ -cladding short-reach UC-4CFs of which cores are arranged in a trapezoidal layout [15] or a linear array [15, 30] have also been fabricated. The theoretical model described here can be extended to such non-square-layout UC-4CFs.

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## Appendix

Eqs. (1a)–(1d) are a set of simultaneous ordinary differential equations. These equations can be converted into a matrix form and the coefficient matrix has three eigenvalues: 0,  $-2(h + g)$ , and  $-4h$ . The eigenvalue  $-2(h + g)$  is doubly degenerate. In general, it is complicated to obtain the eigenvectors corresponding to the degenerate eigenvalue.

Here, to resolve the degeneracy problem, we note that when core 1 is excited at the fiber input end ( $z = 0$ ),  $P_2(z)$

**Table III** Eigenvalues and eigenvectors of Eqs. (14a)–(14c).

$k$	1	2	3
$\gamma_k$	0	$-2(h + g)$	$-4h$
$\mathbf{V}_k$	$\begin{bmatrix} 1 \\ 2 \\ 1 \end{bmatrix}$	$\begin{bmatrix} 1 \\ 0 \\ -1 \end{bmatrix}$	$\begin{bmatrix} 1 \\ -2 \\ 1 \end{bmatrix}$

**Table IV** Eigenvalues and eigenvectors of Eqs. (10a)–(10d).

$k$	1	2	3	4
$\gamma_k$	0	$-2(\tilde{h} + g_1)$	$-2(\tilde{h} + g_2)$	$-4\tilde{h}$
$\mathbf{V}_k$	$\begin{bmatrix} 1 \\ 1 \\ 1 \\ 1 \end{bmatrix}$	$\begin{bmatrix} 1 \\ 0 \\ -1 \\ 0 \end{bmatrix}$	$\begin{bmatrix} 0 \\ 1 \\ 0 \\ -1 \end{bmatrix}$	$\begin{bmatrix} 1 \\ -1 \\ 1 \\ -1 \end{bmatrix}$

and  $P_4(z)$  are equal to each other, and we define  $P_{24}$  as  $P_{24}(z) \equiv P_2(z) + P_4(z)$  and rewrite Eqs. (1a)–(1d) as

$$\frac{dP_1(z)}{dz} = -(2h + g)P_1 + hP_{24} + gP_3, \quad (14a)$$

$$\frac{dP_{24}(z)}{dz} = 2hP_1 - 2hP_{24} + 2hP_3, \quad (14b)$$

$$\frac{dP_3(z)}{dz} = gP_1 + hP_{24} - (2h + g)P_3. \quad (14c)$$

A general solution of Eqs. (14a)–(14c) is given by

$$\begin{bmatrix} P_1(z) \\ P_{24}(z) \\ P_3(z) \end{bmatrix} = \sum_{k=1}^3 C_k e^{\gamma_k z} \mathbf{V}_k, \quad (15)$$

where  $C_k$  ( $k = 1, 2, 3$ ) is the arbitrary constant, and  $\gamma_k$  and  $\mathbf{V}_k$  are the eigenvalue and eigenvector of Eqs. (14a)–(14c), respectively, as shown in Table III. For simplicity, the eigenvectors are not normalized (it is not always necessary to normalize the eigenvectors).

Applying the initial conditions  $P_1(0) = P_{\text{in}}$  and  $P_{24}(0) = P_3(0) = 0$  to Eq. (15), and using  $\mathbf{V}_k$ , we determined the arbitrary constants as

$$C_1 = C_2/2 = C_3 = P_{\text{in}}/4. \quad (16)$$

Then, using  $\gamma_k$ ,  $\mathbf{V}_k$ , and  $C_k$ , and noting  $P_2(z) = P_4(z) = P_{24}(z)/2$ , from Eq. (15) we have the exact solutions derived by Liang *et al.* ( $g \neq 0$ ) [25] and Nakagawa *et al.* ( $g = 0$ ) [24]. When core 3 is excited at the fiber input end, we note that the initial conditions are given by  $P_3(0) = P_{\text{in}}$  and  $P_1(0) = P_{24}(0) = 0$ . When core 2 or 4 is excited at the fiber input end, we note that  $P_1(z) = P_3(z)$  and set  $P_1(z) + P_3(z)$  to  $P_{13}(z)$  in place of  $P_{24}(z)$ .

Table IV shows the eigenvalues and eigenvectors of Eqs. (10a)–(10d) for heterogeneous square-layout UC-4CFs. Eqs. (10a)–(10d) have four eigenvalues, all of which are not degenerate.