

## On the Static Accuracy of Charge-Discharge Units Intended for Electrical Tests of High Capacity Li-ion Batteries

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**Abstract.** Performing of the cycle testing according to the principles of Dynamic Stress Test can significantly reduce the overall time of development and production of batteries, which in turn allows reducing the cost of designing and testing of the spacecraft power systems. Performing of Dynamic Stress Test require special charge-discharge units that allows to perform a full cycle of electrical tests of batteries, including cyclic testing. Providing the required accuracy of measurement and stabilization of certain attributes of Li-ion battery operating modes is one of the problems that arise during the development of such charge-discharge units. The following attributes are of particular interest: charge and discharge currents, discharge powers, battery voltages. Analysis of the charge-discharge unit as a control system allows evaluating the steady-state stabilization error of the required attributes of the developed device. Moreover, using a digital integrator in the control system of the charge-discharge unit allows providing specified values of steady-state stabilization error of required attributes in different test modes.

### 1. Introduction

Li-ion batteries are widely used in spacecraft power systems. Characteristics of Li-ion batteries significantly affect the spacecraft life cycle in orbit. Electrical tests of Li-ion batteries are performed during development and production at the factory, in particular, the battery cycle life is estimated by special designed cycle tests. These tests include multiple charge/discharge cycles in order to reduce total capacity and power of the batteries to a minimum value.

Reducing the time of the cyclic tests can significantly speed up and reduce the cost of designing and testing of spacecraft power systems. In order to reduce the duration of cyclic tests were developed methods of cycle tests for Li-ion batteries based on the method of Dynamic Stress Test (DST) [1–3]. This method involves increasing of charge-discharge currents of the tested battery up to the maximum values, including modes of discharging by constant power. Charging of batteries is typically performed according to the manufacturer's methods, discharging - by the techniques of DST.

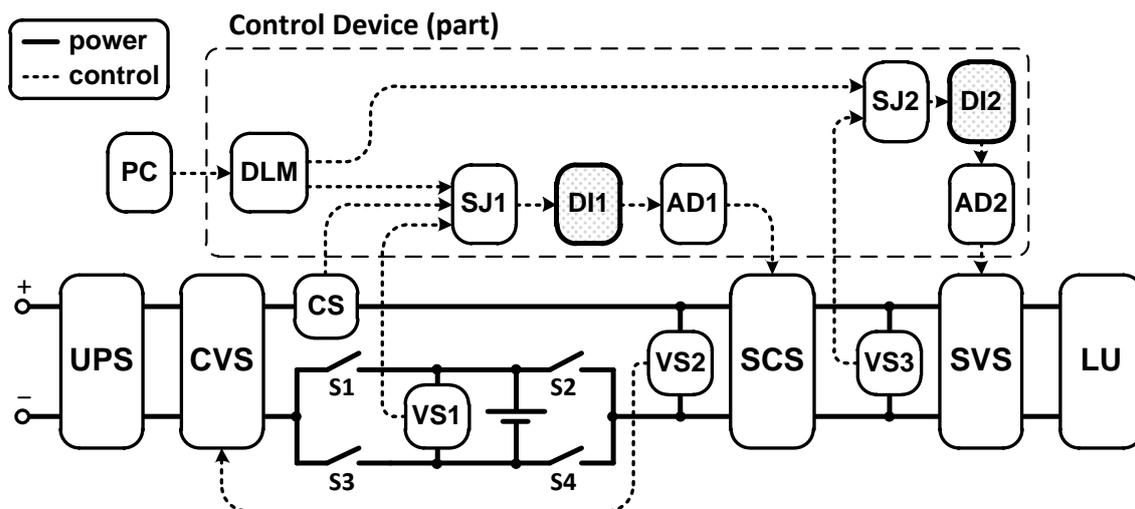
There are many types of charge-discharge units (CDU) for Li-ion batteries [4–8], but these devices are not suitable for testing (including cycle testing) of isolated battery cells that have relatively low voltage and high capacity. There are also devices that allow testing of composite Li-ion batteries [9–10]. The disadvantages of the devices listed above include:



- input current of converter is not high enough: testing of batteries with rated capacity of 90 Ah requires current more than 100 A, which require using of multiple parallel connected devices;
- lack of opportunity to discharge battery to a negative voltage (polarity reversal), which does not allow to study the operation of the battery cells in the emergency situations..

## 2. Charge-discharge unit description

An automated CDU [11–12] (fig.1.) allows performing a full cycle of electrical tests.



**Figure 1.** Block diagram of charge-discharge unit (CDU)

Uninterruptible power supply (UPS) prevents interruption of the tests in the case of emergency AC network shutdown. Charging and discharging modes of the battery are provided by appropriate position of switches S1-S4 and by a controlled voltage stabilizer (CVS). CVS allows ensuring stable charge/discharge mode operation. Excess of energy, consumed by CDU, is dissipated by a load unit (LU).

Necessity of current regulation in the wide range (0..160 A), the need for electrical isolation between switches S1-S4 and LU and complexity of high-frequency power transformer made it necessary to use two more stabilizers switched-mode stabilizer current (SCS) and switched-mode stabilizer voltage (SVS). SCS stabilizes its own input current, SVS stabilizes the SCS output voltage. SCS consist of switched-mode current regulator, which is based on boost converter. SVS consist of switched-mode voltage regulator, which is based on full-bridge buck converter.[13-14]

In order to provide high currents at low battery voltages it is necessary to maintain the SVS input voltage at the relatively high level. For this reason the stabilization of the SCS input voltage is provided by CVS (voltage sensor VS2 forms feedback for CVS). Therefore, regardless of battery voltage and the operating mode (charging or discharging of the battery) CVS stabilizes the SCS input voltage at the level of 4 V. In addition, the stabilization of the SCS input voltage enables to discharge the battery until polarity reversal.

A control device (CD), in accordance with predetermined patterns and the measured values of battery voltages and currents, controls switches S1-S4 (control routes between CD and switches are not shown on the diagram), generates control signals for SCS and SVS, and

provides safe completion of tests if charge of the UPS batteries is low and if voltage in AC network is absent. CD includes digital logic module (DLM) which connected to personal computer (PC) and forms references for SCS and SVS summing junctions (SJ) on which they are added to the signals from the current (CS) and voltage (VS1, VS3) sensors. Control signals for SCS and SVS are filtered by digital integrators (DI) (detailed description will be given below) and amplified by analog drivers (AD).

CDU performs following functions:

- stabilization and regulation of charge and discharge currents;
- stabilization and regulation of discharge power;
- reproduction of electrical and temporal operational modes of battery;
- automatic diagnostics of CDU condition;
- emergency operation protection.

Document [15] gives guidelines for the definition of performance characteristics, guidelines for Li-ion battery cycle testing, and recommended measurement errors for battery parameters which are:

- a)  $\pm 0.1\%$  for voltages;
- b)  $\pm 1\%$  for currents;
- c)  $\pm 0.1\%$  for temporal parameters.

At the same time, document [15] does not establish requirements for the stabilization accuracy of charge and discharge currents and discharge powers. In accordance with the adopted in the design of control systems practice, take attributes stabilization errors equal to measurement errors:

- a)  $\pm 1\%$  for currents;
- b)  $\pm 0.1\%$  for temporal parameters;
- c)  $\pm 1.1\%$  for power.

Providing the required current measurement accuracy is achieved by using high-precision current sensor and by current sensor calibration. Calibration of current sensors performs at the intervals specified in the automatic battery testing program and automatically before startup. Calibration allows reducing to a minimum the statistical component of the current sensor error.

Measurement of the battery voltage and voltages of current sensors is performed using specialized measuring device based on National Instruments modules. Since the electrical isolation of measurement channels is required, NI PXIe-4300 modules are used [16]. They provide 8 channels, which allow simultaneously measuring, and have percentage errors of measurement no more than 0.0245%.

Stabilization and control of the attributes is based on digital signal processing in the single-chip computer. Accuracy of stabilization consists of the sum of the errors:

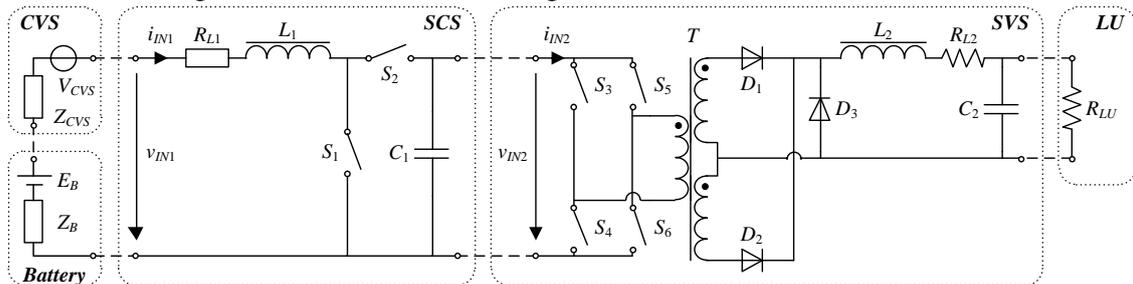
- the current and voltage measurement errors (discussed above);
- errors caused by the analog-digital converter (ADC);
- errors caused by of the reference voltage source;
- errors of attributes stabilization control loops.

Errors of the ADC and reference voltage are minimized by selecting appropriate electronic components. Charge-discharge unit, described by the authors in [11-12], contains a switching regulator of the input current (SRIC) and a switching regulator of the SRIC output voltage (SRSOV).

### 3. Mathematical model

The mathematical model of the power part of CDU is based on the block diagram (fig. 1) and the equivalent circuit (fig. 2). This model allows analyzing the accuracy of CDU static stabilization.

Variations of voltages and currents in circuits are negligible in steady state, which allows omitting the parameters of reactive elements in the mathematical model (inductances and capacitances). A linearized system of steady-state equations (1) describes relations between currents and voltages in the circuit shown in fig. 2:



**Figure 2.** Equivalent electrical circuit of the CDU power part

$$\begin{cases} V_S = V_{CVS} + E_B \\ v_{IN1} = V_S - i_{IN1}(R_{CVS} + R_B) \\ i_{IN1} = I_{IN1}D_I + I_{IN1}d_i + i_{in1}D_I + i_{IN2} \\ v_{IN2} = v_{IN1} - i_{IN1}R_{L1} + V_{IN2}D_I + v_{in2}D_I + V_{IN2}d_i, \\ i_{IN2} = (I_{LU}D_V + i_{lu}D_V + I_{LU}d_v)n \\ v_{LU} = (V_{IN2}D_V + v_{in2}D_V + V_{IN2}d_v)n - i_{LU}R_{L2} \\ i_{LU} = v_{LU}G_{LU} \end{cases} \quad (1)$$

where  $V_S$  – voltage of the source,  $V_{CVS}$  – dc voltage at CVS output,  $E_B$  – battery cell dc voltage,  $v_{IN1}$  – SCS input voltage,  $i_{IN1} = I_{IN1} + i_{in1}$  – total value of SCS input current,  $I_{IN1}$  – equilibrium point value of  $i_{IN1}$ ,  $i_{in1}$  – variation of  $i_{IN1}$ ,  $R_{CVS}$  – CVS resistance,  $R_B$  – battery cell resistance,  $d_I = D_I + d_i$  – total value of SCS duty cycle,  $D_I$  – equilibrium point value of  $d_I$ ,  $d_i$  – variation of  $d_I$ ,  $i_{IN2}$  – SVS input current,  $v_{IN2} = V_{IN2} + v_{in2}$  – total value of SVS input voltage,  $V_{IN2}$  – equilibrium point value of  $v_{IN2}$ ,  $v_{in2}$  – variation of  $v_{IN2}$ ,  $R_{L1}$  – resistance of  $L_1$ ,  $i_{IN2}$  – SVS input current,  $i_{LU} = I_{LU} + i_{lu}$  – total value of LU current,  $I_{LU}$  – equilibrium point value of  $i_{LU}$ ,  $i_{lu}$  – variation of  $i_{LU}$ ,  $d_V = D_V + d_v$  – total value of SVS duty cycle,  $D_V$  – equilibrium point value of  $d_V$ ,  $d_v$  – variation of  $d_V$ ,  $n$  – turns ratio of transformer  $T$ ,  $v_{LU}$  – LU voltage,  $R_{L2}$  – resistance of  $L_2$ ,  $G_{LU}$  – conductance of LU.

The system of equations (2) describes a control system of CDU. Along with (1), it forms a mathematical model of CDU. Block diagram of total model (fig. 3) consists of systems (1) and (2):

$$\begin{cases} d_I = (V_{REFI} - i_{IN1}K_{CS})K_I K_{PWMI} \\ d_V = (v_{IN2}K_{VS} - V_{REFV})K_V K_{PWMV} \end{cases} \quad (2)$$

where  $V_{REFI}$  – reference voltage of SCS feedback loop,  $K_{CS}$  – current sensor gain,  $K_I$  – gain of SCS feedback loop,  $K_{PWMI}$  – gain of SCS pulse width modulator (PWM),  $V_{REFV}$  – reference

voltage of SVS feedback,  $K_{VS}$  – voltage sensor gain,  $K_V$  – gain of SVS feedback loop,  $K_{PWMV}$  – gain of SVS PWM.

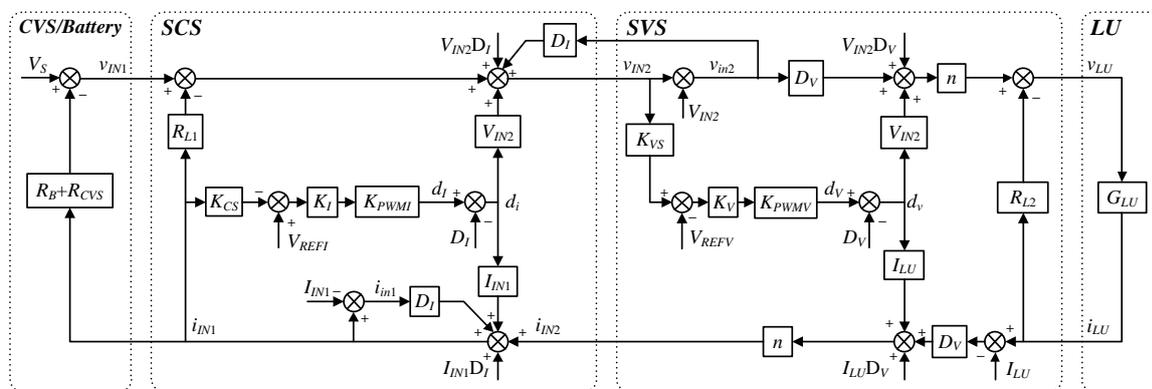
Expressions for the current and voltage with offset that appeared as a result of the presence of steady-state error were obtained from systems (1) and (2):

$$i_{IN1} = f(V_S, V_{REFI}, V_{REFV}, K_I, K_V), \quad (3)$$

$$v_{IN2} = f(V_S, V_{REFI}, V_{REFV}, K_I, K_V). \quad (4)$$

Open-loop transfer functions of SCS and SVS control loops at steady state become SCS gain  $K_{OLI}$  and SVS gain  $K_{OLV}$ , respectively, which are functions of the following variables:

$$K_{OLI} = f(V_S, V_{REFI}, V_{REFV}, K_I, K_V), \quad (5)$$



**Figure 3.** Block diagram of linearized CDU model for steady-state conditions

$$K_{OLV} = f(V_S, V_{REFI}, V_{REFV}, K_I, K_V). \quad (6)$$

Since the CDU input current varies in a wide range of values down to zero, usage of the standard relative error for evaluation of the system accuracy will give to large errors at low currents. Percentage error normalized relative to maximum measurement value (hereinafter – percentage error, designated as  $\delta_M$ ) is more convenient for evaluation purposes [17–18]. The percent error of SCS control loop  $\delta_{MI}$  is used for SCS input current stabilization accuracy estimation:

$$\delta_{MI} = \left| \frac{I_{IN1} - i_{IN1}}{I_{IN1max}} \right|, \quad (7)$$

where  $i_{IN1} = I_{IN1} + i_{in1}$  – total value of SCS input current,  $I_{IN1}$  – equilibrium point value of  $i_{IN1}$ ,  $I_{IN1max}$  – the maximum current of the CDU ( $I_{IN1max} = 160$  A).

SCS percentage error  $\delta_{MI}$  for the CDU was set to 0.5%.

The percent error of SVS control loop  $\delta_{MV}$  is similarly used for SVS input voltage stabilization accuracy estimation:

$$\delta_{MV} = \left| \frac{V_{IN2} - v_{IN2}}{V_{IN2max}} \right|, \quad (8)$$

where  $v_{IN2} = V_{IN2} + v_{in2}$  – total value of SVS input voltage,  $V_{IN2}$  – equilibrium point value of  $v_{IN2}$ ,  $V_{IN2max}$  – the maximum input voltage of SVS.

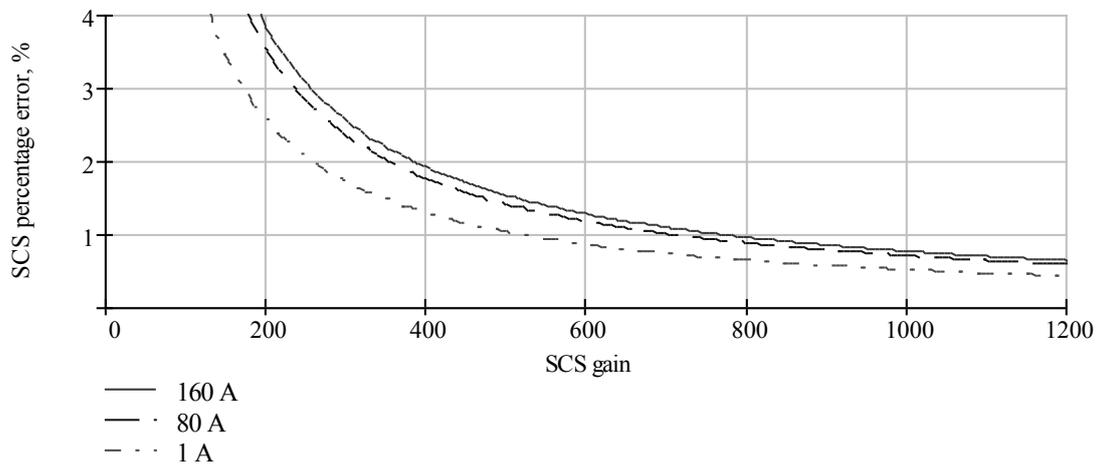
#### 4. Simulation and experimental results

Study the impact of various operational modes of the CDU on the SCS current percentage error allows determining the device parameters that ensure the required accuracy of the CDU input current stabilization. Modes of operation of the CDU are defined by different values of the following quantities: the input current  $I_{IN1}$ , voltage of the source  $V_S$ , SCS gain  $K_{OLI}$  and SVS gain  $K_{OLV}$ .

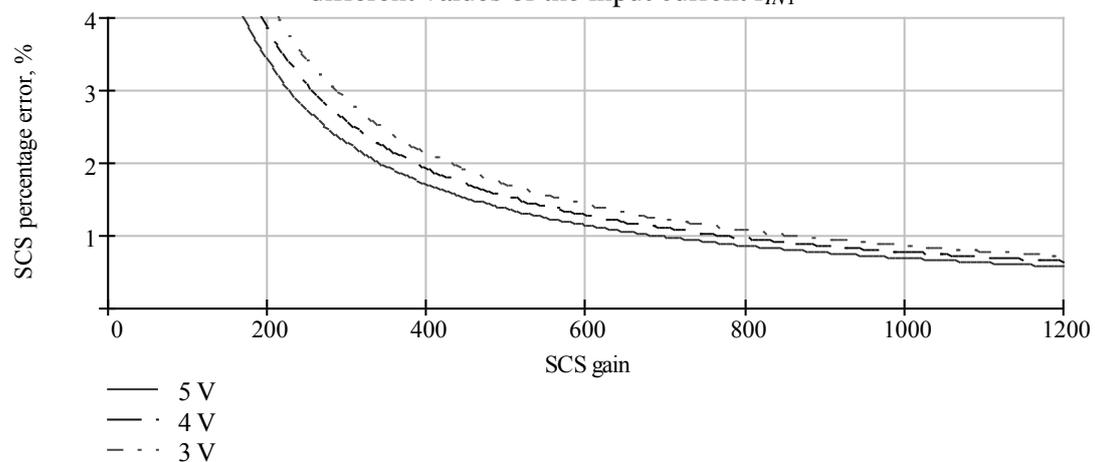
Equations (3)-(8) allow obtaining the implicit dependence of SCS  $\delta_{MI}$  and SVS  $\delta_{MV}$  percentage errors from SCS  $K_{OLI}$  and SVS  $K_{OLV}$  gains (fig.4-9).

Figures 4 and 5 show that with an increase in SCS gain  $K_{OLI}$  value of SCS percentage error  $\delta_{MI}$  decreases, regardless of the action of various destabilizing factors.

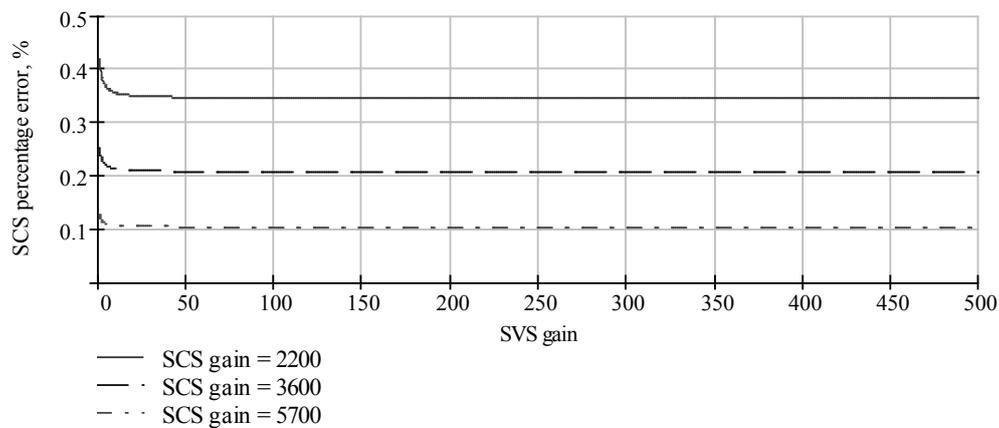
Figure 6 shows that SCS percentage error  $\delta_{MI}$  remains practically constant with increasing of SVS gain  $K_{OLV}$  and decreases with increasing of SCS gain  $K_{OLI}$ .



**Figure 4.** An implicit relation between SCS percentage error  $\delta_{MI}$  error and SCS gain  $K_{OLI}$  for different values of the input current  $I_{IN1}$

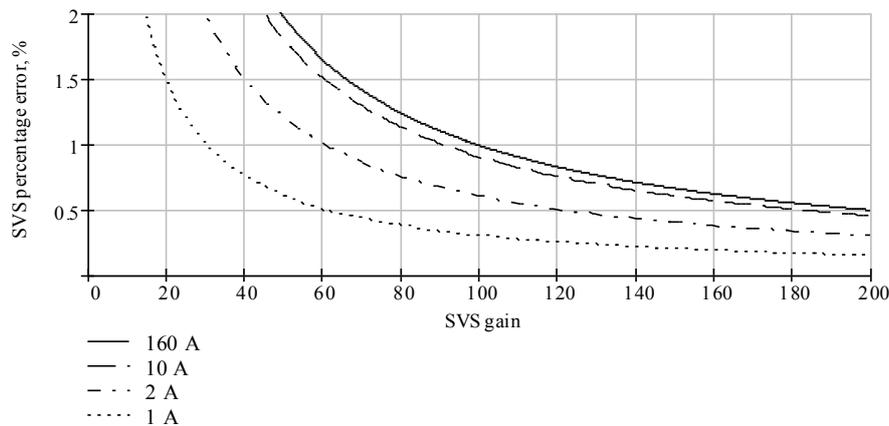


**Figure 5.** An implicit relation between SCS percentage error  $\delta_{MI}$  and SCS gain  $K_{OLI}$  for different values of the input voltage  $V_S$



**Figure 6.** An implicit relation between SCS percentage error  $\delta_{MI}$  and SVS gain  $K_{OLV}$  for different values of SCS gain  $K_{OLI}$

The value of SVS gain  $K_{OLV}$  determines the SVS percentage error  $\delta_{MV}$ . It was established experimentally that for the normal CDU operation a value of SVS percentage error  $\delta_{MV}$  should be 3-5% over the entire range of input voltages and currents. Figures 7-9 show that to ensure this requirement is necessary that the value of SVS gain  $K_{OLV}$  was big enough.

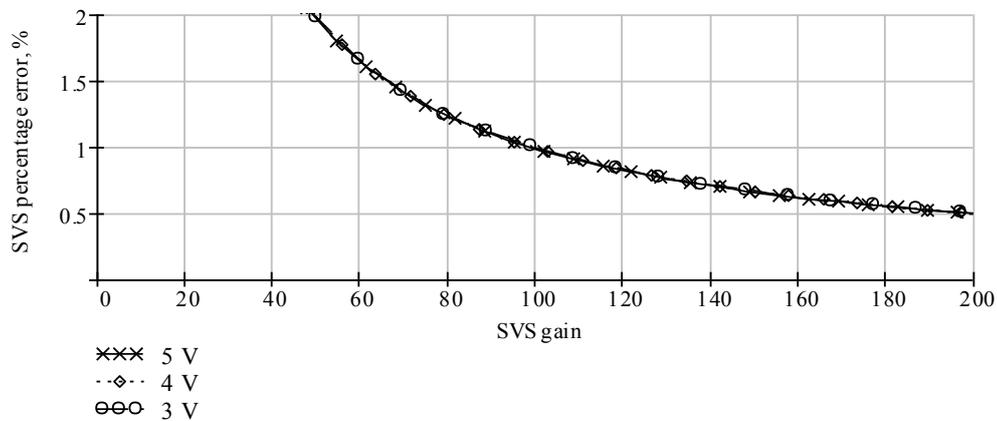


**Figure 7.** An implicit relation between SVS percentage error  $\delta_{MV}$  and SVS gain  $K_{OLV}$  for different values of the input current  $I_{IN1}$

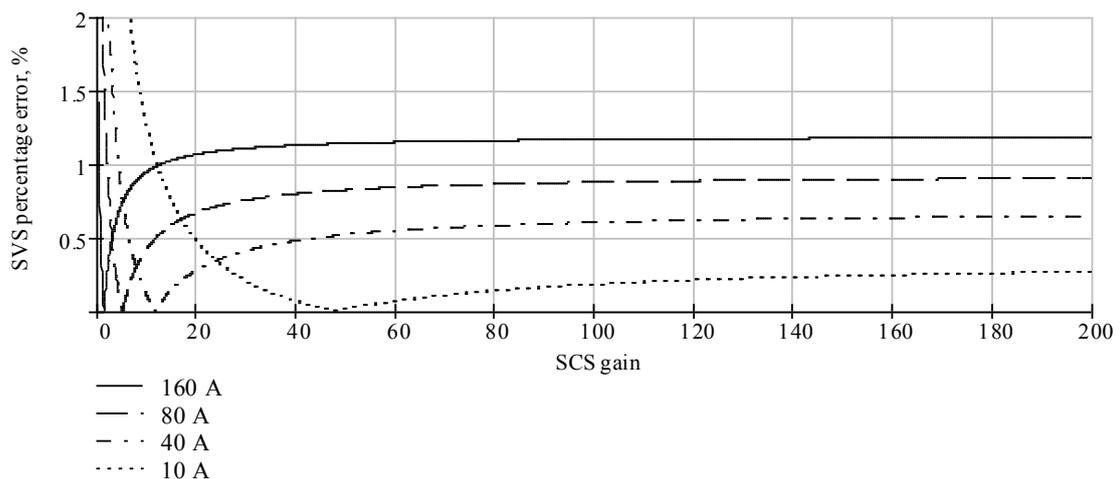
Figures 7 and 8 show that SVS percentage error  $\delta_{MV}$  decreases with increasing SVS gain  $K_{OLV}$  and does not depend on the input voltage  $V_S$ .

Figure 9 shows that SVS percentage error  $\delta_{MV}$  significantly depends on SCS gain  $K_{OLI}$  and the input current  $I_{IN1}$ , which imposes restrictions on the minimum value of SCS gain  $K_{OLI}$ .

It can be concluded from the results of mathematical modeling that SCS gain  $K_{OLI}$  and SVS gain  $K_{OLV}$  are interrelated. It is methodically reasonable to select the value of SVS gain  $K_{OLV}$  first and then to select large enough SCS gain  $K_{OLI}$ .



**Figure 8.** An implicit relation between SVS percentage error  $\delta_{MV}$  and SVS gain  $K_{OLV}$  for different values of the input voltage  $V_S$



**Figure 9.** An implicit relation between SVS percentage error  $\delta_{MV}$  and SCS gain  $K_{OLI}$  for different values of the input current  $I_{IN1}$

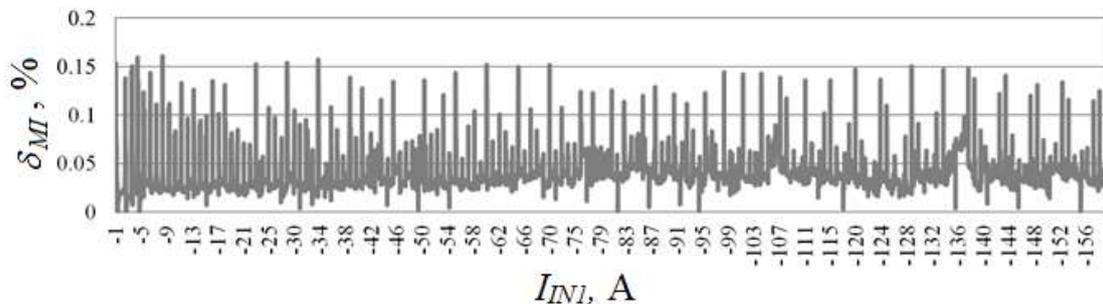
The required accuracy of attribute stabilization was also achieved by increasing system type by least 1 through the introduction of the digital integrator in the SCS control loop [19].

A difference equation of the digital integrator is of the form:

$$E_g[n] = k \cdot E_g[n] + E_g[n-1], \quad (9)$$

where  $n$  – unitless time,  $k$  – integration ratio.

To illustrate the correctness of the developed approach for CDU parameters, experimental studies of SCS percentage error for LISP-85 Li-ion battery [20] were performed (fig. 10).



**Figure 10.** Experimental relation between SCS percentage error  $\delta_{MI}$  and depending on input current  $I_{IN1}$

The experimental results confirm that the SCS percentage error does not exceed 0.2%, which is consistent with the selected design conditions ( $\delta_{MI} = 0.5\%$ ).

## 5. Conclusion

The proposed approach to the selection of the CDU parameters provides the desired percentage errors of required attributes of Li-ion battery operational modes. SCS gain  $K_{OLI}$  and SVS gain  $K_{OLV}$  are interrelated, which can be confirmed by the results of mathematical modelling. For this reason, it makes sense to select the value of SVS gain  $K_{OLV}$  first and then to select large enough SCS gain  $K_{OLI}$ . The required accuracy of attribute stabilization was also achieved by increasing system type by least 1 through the introduction of the digital integrator in the SCS control loop.

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